# Stability Margin of Two-Dimensional Continuous Systems ${ }^{1}$ 

Nikos E.MASTORAKIS
Military Institutions of University Education
Hellenic Naval Academy
Chair of Computer Science
Terma Hatzikyriakou, 18539
Piraeus, GREECE

David H.OWENS
University of Exeter
School of Engineering
Harrison Engineering Building
North Park Road
Exeter EX4 4QF, UNITED KINGDOM
and
Anastasios N. VENETSANOPOULOS
University of Toronto
Dept. of Electr. and Comp. Engineering
Digital Signal and Image Proces.Lab.
10 King's College Road
Toronto, Ontario, CANADA

Abstract:- In this paper, the margin of stability of 2-D (Two-Dimensional) continuous systems is considered. In particular, the definition of the stability margins for a 2-D discrete system is extended to a 2-D continuous one. Two methods to compute the stability margin of 2-D continuous systems are presented. Illustrative examples are provided.

Keywords: Stability, Stability Margin, 2-D systems

[^0]$$
F\left(s_{1}, 1\right) \neq 0, \quad \text { for } \quad \operatorname{Re}\left\{s_{1}\right\} \geq 0 \quad \text { or } \quad s_{1}=\infty
$$

## 1 Introduction

Much interest has been shown during recent years in 2-D continuous systems ([1] $\div[10]$, [17], [30], [41] $\div[45])$ for several reasons: In the design of 2-D and $m-\mathrm{D}(m>2)$ discrete filters, the corresponding analog filters play an important role. In particular, it is possible using appropriate transformations to obtain the desirable 2-D discrete filter from the corresponding analog (2-D) filter [2] $\div[9]$, [41]. On the other hand, in the study of Distributed Parameter Systems (DPS) which are described by Partial Differential Equations (PDEs), each PDE actually corresponds to an $m$-D continuous system. So, for networks which include transmission lines as well as passive lumped elements, for networks containing semiconductor elements, for acoustic filters, the description with 2-dimensional continuous systems is necessary as one can see in [1], [4], [7], [8]. A third reason is the need of the introduction of the 2-D continuous systems theory in Control Systems whose coefficients are functions of parameters as well as in Systems whose inputs and outputs are functions of a time variable and a discrete spatial variable [8], [42] $\div[44]$. For these reasons, there exists an importance of the subject of the $m$-D continuous systems from a practical point of view ([1 $\div 10]$, [17], [23], [30], [41] $\div[45]$ ).

Stability testing of the 2-D and $m$-D $(m>2)$ continuous systems is of much importance [ $1 \div 10$ ], [30]. Let a Linear Shift Invariant 2-D continuous system be described by the transfer function

$$
\begin{equation*}
G\left(s_{1}, s_{2}\right)=\frac{P\left(s_{1}, s_{2}\right)}{F\left(s_{1}, s_{2}\right)} \tag{1}
\end{equation*}
$$

where $P\left(s_{1}, s_{2}\right)$ and $F\left(s_{1}, s_{2}\right)$ are coprime polynomials in the independent complex variables $s_{1}$ and $s_{2}$, It is assumed that there are no nonessential singularities of the second kind on the double imaginary axis, i.e. there are no points $s_{1}, s_{2}$ with $s_{1}=j \omega_{1}$ or $\infty, \quad s_{2}=j \omega_{2}$ or $\infty$ such that $P\left(s_{1}, s_{2}\right)=F\left(s_{1}, s_{2}\right)=0$.

The system (1) is (Hurwitz) stable if and only if
and

$$
\begin{array}{ll}
F\left(j \omega_{1}, s_{2}\right) \neq 0, \quad \text { for }\left(\operatorname{Re}\left\{s_{2}\right\} \geq 0 \text { or } s_{2}=\infty\right) \\
& \text { and }-\infty \leq \omega_{1} \leq \infty \tag{2.2}
\end{array}
$$

Additionally, the polynomial $F\left(s_{1}, s_{2}\right)$ is said to be Hurwitz Polynomial if and only if (2.1) and (2.2) are fulfilled. Condition (2.1) is relatively easy to check using any 1-D stability test. Checking condition (2.2) is a more difficult task.

There exist several algebraic methods for testing the stability of 2-D continuous systems or, equivalently, checking the Hurwitz character of 2-D polynomials [30], [28], [29]. Among them are the table form as advanced by Siljak [13], the determinant method (Anderson-Jury [14]), the inner method [16], [21] as advanced in [24], and the Zeheb-Walach method [12]. Another somewhat tedious approach is the method of the bilinear transformation. However, as pointed out by Goodman [20] and Jury and Bauer [22] the bilinear transformation can cause some difficulties in the stability tests because of the presence of nonessential singularities of the second kind and the value of the function at infinity [18], [19].

In the study of 2-D continuous systems, we are interested not only in whether the system is stable but also whether the system will remain stable in the presence of system parameter deviations.

For this reason and analogously to 2-D discrete systems [31], the following definition is introduced:
Definition 1: Given a 2-D continuous system described by the transfer function (1), we call stability margin $\sigma_{1}$ the greater non positive real number for which $F\left(s_{1}+\sigma_{1}, s_{2}\right)$ is a Hurwitz Polynomial.

Similarly the following two definitions are stated.
Definition 2: Given a 2-D continuous system described by the transfer function (1), we call stability margin $\sigma_{2}$ the greater non positive real number for which $F\left(s_{1}, s_{2}+\sigma_{2}\right)$ is a Hurwitz Polynomial.

Definition 3: Given a 2-D continuous system described by the transfer function (1), we call stability margin $\sigma$ the greater non positive real number for which $F\left(s_{1}+\sigma, s_{2}+\sigma\right)$ is a Hurwitz Polynomial.

Note that the special case where the stable system has nonessential singularities of the second kind for some $s_{1}=j \omega_{1}$ or $\infty$ and $s_{2}=j \omega_{2}$ or $\infty$ is excluded, since all three stability margins will be zero.

In 2-D discrete systems these definitions were given in [31], while several methods for the evaluation of stability margins already exist [31] $\div$ [38]. In this paper, the analogous methods in 2-D continuous case are derived the applicability and effectiveness of which are illustrated by some examples.

## 2 Computation of the stability margins for 2-D continuous systems

## A. Hermite Matrix Method

In this paragraph, a method of computing the stability margin of 2-D continuous systems is presented. The method is based on checking the positive definiteness of the Hermite matrix of the characteristic polynomial $F\left(s_{1}, s_{2}\right)$ of a stable system described by (1). For a 2-D continuous system we recall that the Hermite matrix $H_{1}\left(\omega_{1}\right)$ associated with $F\left(j \omega_{1}, s_{2}\right)$ is positive definite where $-\infty \leq \omega_{1} \leq \infty$ and $\operatorname{Re}\left\{s_{2}\right\} \geq 0$ or $s_{2}=\infty$ [30].

Considering the Hermite matrix $H_{1}\left(\omega_{1}, \sigma_{1}\right)$ associated with $F\left(j \omega_{1}+\sigma_{1}, s_{2}\right)$, we obtain that in the "limit point" of the maximum value of $\sigma_{1}$ the Hermite matrix $H_{1}\left(\omega_{1}, \sigma_{1}\right)$ will be positive semidefinite $\quad\left(-\infty \leq \omega_{1} \leq \infty\right.$, $\operatorname{Re}\left\{s_{2}\right\} \geq 0$ or $s_{2}=\infty$ ). This implies that for this point the matrix $H_{1}\left(\omega_{1}, \sigma_{1}\right)$ is singular. Thus the computation of $\sigma_{1}$ can be achieved by solving the following optimization problem

$$
\begin{equation*}
\max \sigma_{1} \tag{3}
\end{equation*}
$$

$$
\sigma_{1} \leq 0
$$

under the constraint

$$
\begin{equation*}
\operatorname{det} H_{1}\left(\omega_{1}, \sigma_{1}\right)=0 \tag{4}
\end{equation*}
$$

where $H_{1}\left(\omega_{1}, \sigma_{1}\right)$ is the Hermitian matrix associated with $F\left(j \omega_{1}+\sigma_{1}, s_{2}\right)$. By interchanging the roles of the variables $s_{1}$ and $s_{2}$, a completely analogous method for the computation of $\sigma_{2}$ is obtained. Moreover, for the computation of $\sigma$ we demand

$$
\begin{align*}
& \max \sigma  \tag{5}\\
& \sigma \leq 0
\end{align*}
$$

under the constraint

$$
\begin{equation*}
\operatorname{det} H\left(\omega_{1}, \sigma\right)=0 \tag{6}
\end{equation*}
$$

where $H\left(\omega_{1}, \sigma\right)$ is the Hermitian matrix associated with $F\left(j \omega_{1}+\sigma, s_{2}+\sigma\right)$. The following example illustrates the implementation of this method.

Example 1: Consider the characteristic polynomial of a 2-D (continuous) system

$$
\begin{equation*}
F\left(s_{1}, s_{2}\right)=7+s_{1}+2 s_{2}+2 s_{1} s_{2} \tag{7}
\end{equation*}
$$

It is always assumed that the corresponding 2-D system has no nonessential singularities of the second kind. Obviously, condition (2.1) holds while condition (2.2) can be easily checked via the positive definiteness of the Hermitian matrix which is $H_{1}\left(\omega_{1}\right)=14+2 \omega_{1}^{2}$. Therefore $F\left(s_{1}, s_{2}\right)$ is a Hurwitz Polynomial. For the computation of the stability margin $\sigma_{1}$, one forms the Hermitian determinant (i.e. the determinant of the Hermitian matrix) of $F\left(j \omega_{1}+\sigma_{1}, s_{2}\right)$. This is

$$
\begin{equation*}
H_{1}\left(\omega_{1}, \sigma_{1}\right)=\left(2+2 \sigma_{1}\right)\left(7+\sigma_{1}\right)+2 \omega_{1}^{2} \tag{8}
\end{equation*}
$$

The maximum value of $\sigma_{1}\left(\sigma_{1} \leq 0\right)$ for which $H_{1}\left(\omega_{1}, \sigma_{1}\right)$ is nullified for some $\omega_{1}$ is obviously -1 . Therefore the stability margin $\sigma_{1}$ is -1 . Quite analogously one finds that $\sigma_{2}=-\frac{1}{2}$. In order to evaluate the stability margin $\sigma$, we form the Hermitian determinant associated with $F\left(j \omega_{1}+\sigma, s_{2}+\sigma\right)$. This can be found as

$$
\begin{equation*}
H\left(\omega_{1}, \sigma\right)=(2+2 \sigma)\left(7+3 \sigma+2 \sigma^{2}\right)+2 \omega_{1}^{2}(1+2 \sigma) \tag{9}
\end{equation*}
$$

We obtain that for $\sigma \rightarrow-\frac{1^{+}}{2}, H\left(\omega_{1}, \sigma\right)$ is positive, but for $\sigma \rightarrow-\frac{1^{-}}{2}$ and $\omega_{1} \rightarrow \infty, H\left(\omega_{1}, \sigma\right)$ is negative. So, if $\sigma \rightarrow-\frac{1^{-}}{2}$ and one allows $\omega_{1}=\sqrt{-\frac{(2+2 \sigma)\left(7+3 \sigma+2 \sigma^{2}\right)}{2(1+2 \sigma)}}(\longrightarrow \infty)$, then $H\left(\omega_{1}, \sigma\right)=0$. Therefore $\sigma=-\frac{1}{2}$.

Example 2: Consider the general first order characteristic polynomial of a 2-D (continuous) system

$$
\begin{equation*}
F\left(s_{1}, s_{2}\right)=1+a s_{1}+b s_{2}+c s_{1} s_{2} \tag{10}
\end{equation*}
$$

where $a, b, c>0$. By verifying the same conditions as in Example 1, one can easily see that this a Hurwitz Polynomial. Similarly one finds

$$
\begin{equation*}
H_{1}\left(\omega_{1}, \sigma_{1}\right)=\left(b+c \cdot \sigma_{1}\right)\left(1+a \cdot \sigma_{1}\right)+a \cdot c \cdot \omega_{1}^{2} \tag{11}
\end{equation*}
$$

So, the maximum non-positive value of $\sigma_{1}$ for which $H_{1}\left(\omega_{1}, \sigma_{1}\right)$ is nullified for some $\omega_{1}$ is:

$$
\begin{equation*}
\sigma_{1}=\max \left\{-\frac{b}{c},-\frac{1}{a}\right\} \tag{12}
\end{equation*}
$$

Consequently, interchanging the variables $s_{1}$ and $s_{2}$ one evaluates

$$
\begin{equation*}
\sigma_{2}=\max \left\{-\frac{a}{c},-\frac{1}{b}\right\} \tag{13}
\end{equation*}
$$

To obtain $\sigma$, we form the Hermitian determinant associated with $F\left(j \omega_{1}+\sigma, s_{2}+\sigma\right)$.
$H\left(\omega_{1}, \sigma\right)=(b+c \cdot \sigma)\left(1+(a+b) \cdot \sigma+c \cdot \sigma^{2}\right)+\omega_{1}^{2} \cdot c \cdot(a+c \cdot \sigma)$

We assert that $\sigma$ is the maximum of the real roots of the polynomials (in $\sigma$ )
$(b+c \cdot \sigma),\left(1+(a+b) \cdot \sigma+c \cdot \sigma^{2}\right),(a+c \cdot \sigma) . \quad$ То prove this, we suppose that $-\frac{a}{c}$ is greater than the (real) roots of the polynomials $(b+c \cdot \sigma),\left(1+(a+b) \cdot \sigma+c \cdot \sigma^{2}\right)$. Then if $\sigma \rightarrow-\frac{a^{-}}{c}$ and
$\omega_{1}=\sqrt{-\frac{(b+c \sigma)\left(1+(a+b) \sigma+c \sigma^{2}\right)}{c(a+c \sigma)}},(\rightarrow \infty$
since $(b+c \sigma)\left(1+(a+b) \sigma+c \sigma^{2}\right)>0$ for $\sigma \rightarrow-\frac{a^{-}}{c}$ $(a, b, c>0)$ ) we obtain that $H\left(\omega_{1}, \sigma\right)=0$. On the other hand, if one root $\sigma_{r}$ of the polynomials $(b+c \cdot \sigma),\left(1+(a+b) \cdot \sigma+c \cdot \sigma^{2}\right)$ is greater than $-\frac{a}{c}$ then for $\sigma=\sigma_{r}$ and $\omega_{1}=0$ we find $H\left(\omega_{1}, \sigma\right)=0$. Hence,
$\sigma=\left\{\begin{array}{l}\max \left(\frac{-(a+b)+\sqrt{(a+b)^{2}-4 c}}{2},-\frac{b}{c},-\frac{a}{c}\right) \text { if }(a- \\ \max \left(-\frac{b}{c},-\frac{a}{c}\right)\end{array}\right.$

It is worth noticing the symmetry between $a$ and $b$.
$\omega_{2}=+\mathrm{oo}$

$\mathrm{O}_{2}=-\mathrm{OO}$

Fig.1: Nyquist curve of the complex variable

$$
s_{2}=j \omega_{2}
$$



Fig.2: A geometrical interpretation of Equation (17)

Differentiating (17) along the considered branch of $s_{1}$, one obtains

$$
\begin{equation*}
\frac{\partial F\left(s_{1}, j \omega_{2}\right)}{\partial s_{1}} d s_{1}+\frac{\partial F\left(s_{1}, j \omega_{2}\right)}{\partial \omega_{2}} d \omega_{2}=0 \tag{18}
\end{equation*}
$$

Since the solution of the (1-Dimensional) Eq. (17) with respect to $s_{1}$ is always possible, we have that $\frac{\partial F\left(s_{1}, j \omega_{2}\right)}{\partial s_{1}} \neq 0$ [26], [27]. Therefore, (18) renders,

$$
\begin{equation*}
\frac{\partial s_{1}}{\partial \omega_{2}}=-\frac{\frac{\partial F\left(s_{1}, j \omega_{2}\right)}{\partial \omega_{2}}}{\frac{\partial F\left(s_{1}, j \omega_{2}\right)}{\partial s_{1}}} \tag{19}
\end{equation*}
$$

Equation (19) guarantees the existence of the (geometrical) tangent of all the branches of (17) at any point. Moreover, at the closest point of all these branches to the imaginary axis, one observes that the tangent will be parallel to the imaginary axis (Fig.2). Therefore a simple necessary condition for the stability margin $\sigma_{1}$ will be

$$
\begin{equation*}
\operatorname{Arg}\left[\frac{\partial s_{1}}{\partial \omega_{2}}\right]= \pm \frac{\pi}{2} \tag{20}
\end{equation*}
$$

Taking into account that $s_{2}=j \omega_{2}$, one rewrites (19) as

$$
\begin{equation*}
\frac{\partial s_{1}}{\partial \omega_{2}}=-\left.j \frac{\frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{2}}}{\frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{1}}}\right|_{s_{2}=j \omega_{2}} \tag{21}
\end{equation*}
$$

Therefore

$$
\begin{equation*}
\operatorname{Arg}\left[\left.\frac{\frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{2}}}{\frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{1}}}\right|_{s_{2}=j \omega_{2}}\right]=0 \text { or } \pi \tag{22}
\end{equation*}
$$

It is more convenient to write Equation (22) in the form

$$
\begin{equation*}
\left.\left(\frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{1}} \times \frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{2}}\right)\right|_{s_{2}=j \omega_{2}}=0 \tag{23}
\end{equation*}
$$

where the outer product ( $\times$ ) of the complex numbers $x_{1}+j y_{1}$ and $x_{2}+j y_{2}$ is defined as the algebraic value of the outer product of the corresponding vectors $\left(x_{1}, y_{1}\right)$ and $\left(x_{2}, y_{2}\right)$ i.e. $x_{1} y_{2}-x_{2} y_{1}$. We could also have written Equation (23) in a more general form

$$
\begin{equation*}
\left.\left(\frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{1}} / / \frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{2}}\right)\right|_{s_{2}=j \omega_{2}} \tag{24}
\end{equation*}
$$

where // denotes the parallelness of the corresponding vectors.
Thus, the stability margin $\sigma_{1}$ is evaluated by the solution of (17) and (23) (or (24)). These two equations can readily be put in the form of three polynomial equations in the three independent variables $\quad x, y, \omega_{2}$, where $\quad x=\operatorname{Re}\left\{s_{1}\right\} \quad$ and $y=\operatorname{Im}\left\{s_{1}\right\}$. Their common solution(s) can be found, if desired, using the resultant method of these polynomials [16], [39]. In the sequel, one must select as $\sigma_{1}$ the maximum $x$ of all the solutions $\left(x, y, \omega_{2}\right)$.

A similar method can also be formulated to derive the stability margin $\sigma_{2}$. For the stability margin $\sigma$, instead of (17) and (23) we get the equations

$$
\begin{equation*}
\left.F\left(s_{1}, s_{2}\right)\right|_{\substack{s_{1}=x+j y \\ s_{2}=x+j \omega_{2}}}=0 \tag{25}
\end{equation*}
$$

and

$$
\begin{equation*}
\left.\left(\frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{1}} \times \frac{\partial F\left(s_{1}, s_{2}\right)}{\partial s_{2}}\right)\right|_{\substack{s_{1}=x+j y \\ s_{2}=x+j \omega_{2}}}=0 \tag{26}
\end{equation*}
$$

These equations can also be put in the form of three polynomial equations in $x, y, \omega_{2}$. Their common solution(s) can be derived if desired, from the resultant method. Furthermore, the stability margin $\sigma$ is evaluated as the maximum $x$ of all the solutions ( $x, y, \omega_{2}$ ).

Example 3: To demonstrate the method, we consider the same characteristic polynomial as in Example 2.

$$
F\left(s_{1}, s_{2}\right)=1+a s_{1}+b s_{2}+c s_{1} s_{2}
$$

where $a, b, c>0$. Eq.(17) and (23) yield

$$
\begin{align*}
1+a x-c y \omega_{2} & =0  \tag{27}\\
a y+(b+c x) \omega_{2} & =0  \tag{28}\\
a c y-c \omega_{2}(b+c x) & =0 \tag{29}
\end{align*}
$$

where $x=\operatorname{Re}\left\{s_{1}\right\}$ and $y=\operatorname{Im}\left\{s_{1}\right\}$.
If $\omega_{2} \neq \infty$, then solving (28) with respect to $(b+c x) \omega_{2}$ and substituting into (29), one finds $y=0$. Then, from (27), $x=-\frac{1}{a}$. If $\omega_{2} \rightarrow \infty$, (28) implies that $y \rightarrow \infty$ or $b+c x=0$. However, if $y \rightarrow \infty$, from (27) we obtain $x \rightarrow \infty$ which is not accepted. Therefore $b+c x=0$ or $x=-\frac{b}{c}$ and $y=\frac{c-a b}{c^{2} \omega_{2}} \xrightarrow{\omega_{2} \rightarrow \infty} 0$. In this case, (28) and (29)
hold considering limits in their left hand side. Therefore

$$
\begin{equation*}
\sigma_{1}=\max \left\{-\frac{b}{c},-\frac{1}{a}\right\} \tag{30}
\end{equation*}
$$

Similarly, interchanging the variables $s_{1}$ and $s_{2}$

$$
\begin{equation*}
\sigma_{2}=\max \left\{-\frac{a}{c},-\frac{1}{b}\right\} \tag{31}
\end{equation*}
$$

For the stability margin $\sigma$, Eq.(25) and (26) render

$$
\begin{align*}
1+(a+b) x+c x^{2}-c \omega_{2} y & =0  \tag{32}\\
(a+c x) y+(b+c x) \omega_{2} & =0  \tag{33}\\
(a+c x) c y-c \omega_{2}(b+c x) & =0 \tag{34}
\end{align*}
$$

From (32), $y=\frac{1+(a+b) x+c x^{2}}{c \omega_{2}}$. Multiplying (33) by $c \omega_{2}$ and substituting the term $c \omega_{2} y$ from (32), we get

$$
\begin{equation*}
(a+c x)\left(1+(a+b) x+c x^{2}\right)+(b+c x) c \omega_{2}^{2}=0 \tag{35}
\end{equation*}
$$

Similarly, (34) yields
$(a+c x)\left(1+(a+b) x+c x^{2}\right)-(b+c x) c \omega_{2}^{2}=0$

Obviously, from (35) and (36) if $\omega_{2} \neq \infty$, one obtains $(a+c x)\left(1+(a+b) x+c x^{2}\right)=0$ as well as $\omega_{2}=0$. On the other side, if $-\frac{b}{c}$ is greater than any real root of $(a+c x)\left(1+(a+b) x+c x^{2}\right)$ and if $\omega_{2} \rightarrow \infty$, from (35), one finds $x \rightarrow-\frac{b^{-}}{c}$. So, if we let $\quad x \rightarrow-\frac{b^{-}}{c}, \quad$ then $\quad$ we obtain
$\omega_{2}=\sqrt{-\frac{(a+c x)\left(1+(a+b) x+c x^{2}\right)}{b+c x}} \rightarrow \infty$. In this case, (33) and (34) hold considering limits in the left hand side. Evidently, $\sigma$ is the maximum of all $x$, i.e.
the maximum of $-\frac{b}{c}$ and the real roots of $(a+c x)\left(1+(a+b) x+c x^{2}\right)$. Thus, the stability margin $\sigma$ is given by Eq.(15).

Remark 1: All the above results are extended to the $m$-D ( $m>2$ ) continuous case after simple modifications.
Remark 2: An interesting generalization of the Definitions of $\sigma_{1}, \sigma_{2}, \sigma$ could be the following: Definition of the stability margin $\sigma$ with weights $k, \lambda \quad(k+\lambda=1, k \geq 0, \lambda \geq 0)$ : Given a 2-D continuous system described by the transfer function (1), we call stability margin $\sigma$ the greater non positive real number for which $F\left(s_{1}+k \sigma, s_{2}+\lambda \sigma\right)$ is a Hurwitz Polynomial Modifying anyone of the above methods, one can easily derive two appropriate algorithms for evaluating the stability margin $\sigma$ with weights $k, \lambda$.

## 3 Conclusions

In many engineering applications (study of distributed parameter systems, design of 2-D discrete filters via appropriate transformations of 2-D continuous ones, study of automatic control systems whose coefficients are functions of parameters and study of systems whose inputs and outputs are functions of a time variable and a discrete spatial variable) the class of 2-D continuous systems has been proved as a very useful class. In the study of 2D continuous systems, we are interested not only in whether the system is stable but also how far is the system from the region of instability. A measure of this is the so called stability margin.

Two methods for computing the stability margin of 2-D continuous systems have been proposed in this paper. The first method is based on a constrained optimization of a real non-positive parameter. The second method is based on a geometrical consideration of the whole problem. Some examples
are given which also show the equivalence of the methods.

One could obtain that the first method is simpler and easy to understand but an optimization problem is needed to be solved. Because of the use of optimization, this method does not always yield an exact value. From this point of view, the geometrical method seems to be better, even if we have to select the solution with the $\max (x)$.

## References

[1] H.G.Ansell, "On certain two-variables generalization of circuit theory with applications to networks of transmission lines and lumped reactances", IEEE Trans.Circuit Theory, Vol.11, pp.214-223, June 1964.
[2] S.Erfani and B.Peikari, "Digital Design of Two-Dimensional LC Structures", IEEE Trans. Circ. Syst., vol. CAS-28, No.1, pp. 75-77, Jan. 1981.
[3] S.Erfani, M.Ahmadi and V.Ramachandran, "Modified realization technique for digital filters derived from analog counterparts", J.of the Franklin Institute, Vol.322, No.4, pp.221-228, Octob. 1986.
[4] M.N.S. Swamy and Harnath C.Reddy, "TwoVariable Analog Ladders with Applications", in Multidimensional Systems (Techniques and Applications), Marcel Dekker, Inc, New York 1986, edited by Spyros G. Tzafestas.
[5] Q.Liu and L.T.Bruton, "Design of 3-D planar and beam recursive digital filters using spectral transformations", IEEE Trans. Circ. Syst., vol. CAS-36, No.3, pp. 365-374, March 1989.
[6] J.H.Lodge and M.M.Fahmy, "The bilinear transformation of two-dimensional state -space systems," IEEE Trans. Acoust. Speech and Signal Processing, vol. ASSP-30, pp. 500-502, June, 1982.
[7] F.L.Lewis, W.Marszalek and B.G.Mertzios, "Walsh function analysis of 2-D generalized continuous systems, IEEE Trans. Autom. Control, Vol.AC-35, pp.1140-1144, 1990.
[8] E.W.Kamen, "Stabilization of Linear SpatiallyDistributed Continuous-Time and DiscreteTime Systems" in Multidimensional Systems Theory, Progress, Directions and Open Problems in Multidimensional Systems, D.

Reidel Publishing Company, Dordrecht, Holland 1985, edited by N K.Bose.
[9] N.E.Mastorakis, "A continuous model for 2Dimensional LSI Systems", Proceedings of the 1st International Conference on Circuits, Systems and Computers (CSC'96), Hellenic Naval Academy, Volume 1, pp.19-35, Piraeus, Greece, July 15-17, 1996.
[10] S.Chakrabarti, B.B.Bhattacharyya and M.N.S.Swamy, "Approximation of two-variable filter specifications in analog domain, IEEE Trans. Circ. Syst., vol. CAS-24, pp.378-388, July 1977.
[11] J.P.Guiver and N.K.Bose, "On Test for ZeroSets of Multivariate Polynomials in Noncompact Polydomains", Proceedings of the IEEE, Vol.69, No.4, pp.467-469, April 1981.
[12] E.Zeheb and E.Walach, "Zero Sets of Multiparameter Functions and Stability of Multidimensional Systems", IEEE Trans. Acoust. Speech \& Signal Proc., vol. ASSP-29, No.2, pp.197-206, April 1981.
[13] D.D.Siljak, "Stability Criteria for TwoVariables Polynomials", IEEE Trans. Circ. Syst., vol. CAS-22, No.3, pp.185-189, March 1975.
[14] B.D.O.Anderson and E.I.Jury, "Stability Test for Two-Dimensional Recursive Filters", IEEE Trans. Audio \& Electroacoustics, Vol. AU-21, No.4, pp.366-372, August 1973.
[15] N.K.Bose and E.I.Jury, "Positivity and Stability tests for multidimensional filters (discrete and continuous), IEEE Trans. Acooust. Speech Signal Process., vol.ASSP-22, No.3, pp.174180, June 1974.
[16] E.I.Jury, "Theory and Applications of the inners", Procedings of the IEEE, Vol.63, No.7, pp.1044-1068, July 1975.
[17] T.Kaczorek, Two-Dimensional Linear Systems, Springer-Verlag, Lecture Notes in Control and Information Sciences, Berlin-Heidelberg, 1985.
[18] D.Goodman, "Some Stability Properties of Two-Dimensional Linear Shift-Invariant Digital Filters, IEEE Transactions on Circ.\& Syst., Vol.CAS-24, No.4, pp201-208,April 1977
[19] L.M.Roytman, M.N.S.Swamy and G.Eichman, "BIBO Stability in the presence of Nonessential Singularities of the second kind in 2-D Digital Filters", IEEE Transactions on Circ.\& Syst., Vol.CAS-34, No.1, pp.60-72,April 1987.
[20] D.Goodman, "Some difficulties with double bilinear transformation in 2-D filter design", IEEE Proceedings, vol.66, p.983, Aug. 1978.
[21] E.I.Jury and M.Mansour, "A note on new Innermatrix for stability, Procedings of the IEEE, Vol.69, pp.1579-1580, Dec. 1981.
[22] E.I.Jury and P.Bauer, "On the stability of twodimensional continuous systems", IEEE Trans.Circ.Syst., CAS-35, pp.1487-1500, Dec. 1988.
[23] N.E.Mastorakis, N.J.Theodorou and S.G.Tzafestas, "A General Factorization method for multivariable polynomials", Multidimensional Systems and Signal Processing, 5, pp.151-178, 1994.
[24] A.J.Kanellakis, S.G.Tzafestas and N.J.Theodorou, "Stability Tests for 2-D systems Using the Schwarz form and inners determinants", IEEE Trans.Circ.Syst., CAS-38, Vol.9, pp.1071-1077, Sept. 1991.
[25] N.E.Mastorakis, "Comments Concerning Multidimensional Polynomials' Properties" IEEE Trans.Autom. Control, Vol. AC-41, No.2, p.260, Feb. 1996.
[26] Tom M. Apostol, Mathematical Analysis, Addison-Wesley Publishing Company, 1974, (Fifth Printing 1981).
[27] Louis Brand, Advanced Calculus, John Wiley and Sons, 1955.
[28] P.K.Rajan and C.Reddy, "Modified Ansell's method and testing of very strict Hurwitz Polynomials" IEEE Trans. Circ. Syst., vol.CAS-34, pp.559-561, June 1987.
[29] P.Delsarte, Y.V.Genin and Y.Kamp, "Two variable stability criteria", in Proc. ISCA, Tokyo, pp.495-498, 1979.
[30] E.I.Jury, "Stability of Multidimensional Systems and Related Problems", in Multidimensional Systems (Techniques and Applications), Marcel Dekker, Inc, New York 1986, edited by Spyros G. Tzafestas.
[31] P.Agathoklis, E.I.Jury and M.Mansour, "The margin of stability of 2-D linear discrete systems", IEEE Trans. Acoust. Speech Signal Process., Vol. ASSP-30, No.6, pp.869-873, Dec. 1982.
[32] E.Walach and E.Zeheb, " $N$-Dimensional stability margins computation and a variable transformation", IEEE Trans. Acoust. Speech Signal Process., Vol. ASSP-30, No.6, pp.887893, Dec. 1982.
[33] W.-S. Lu and C.B.Lee, "Stability analysis for 2-D systems via a Lyapunov approach" IEEE Trans. Circ. Syst., vol.CAS-32, pp.61-68, Feb. 1985.
[34] L.M.Roytman, M.N.S.Swamy and G.Eichmann, "An efficient numerical scheme to compute 2-D stability thresholds", IEEE Trans. Circ. Syst., vol.CAS-34, No.3, pp.322-324, March 1987.
[35] M.Wang, E.B.Lee and D.Booley, "A simple method to determine the stability and the margin of stability of 2-D recursive filters". IEEE Trans. Circ. Syst. -I: Fundamental Theory and Applications, vol.39, No.3, pp.237-239, March 1992.
[36] D.Hertz and E.Zeheb, "Simplifications in Multidimensional stability margin computations, IEEE Trans. Acoust. Speech Signal Process., Vol. ASSP-35, No.4, pp.566568, April1987.
[37] W.S.Lu, A.Antoniou and P.Agathoklis, "Stability of 2-D Digital Filters under parameter variations", IEEE Trans. Circ. Syst., vol.CAS33, No.5, pp.476-482, May 1988.
[38] P.Agathoklis, "Lower bounds for the stability margin of discrete 2-D systems based on the Lyapunov equation", IEEE Trans. Circ. Syst., vol.CAS-35, No.6, pp.745-749, June 1988.
[39] T.A.Bickart and E.I.Jury, "Real Polynomials: Nonnegativity and positivity", IEEE Trans. Circ.\& Syst., vol.CAS-26, pp.676-683, Sept. 1978.
[40] R.P.Roesser, "A discrete state-space model for image processing," IEEE Trans. Automat. Contr., vol. AC-20, pp. 1-10, Feb. 1975.
[41] W.-S. Lu and A. Antoniou, Two-Dimensional Digital Filters, Marcel Dekker, 1992.
[42] E.W.Kamen and P.P.Khargonekar, "On the Control of Linear Systems whose coefficients are functions of parameters", IEEE Trans. Automat. Contr., vol. AC-27, pp. 627-638, 1982.
[43] C.I.Byrnes, "On the stabilizability of Linear Control Systems Depending on Parameters", Proceedings of 18th IEEE Conference on Decision and Control, Ft. Lauderdale, pp.233236, 1979.
[44] A.S.Morse, "Ring Models for Delay Differential Systems", Automatica, Vol.12, pp.529-531, 1976.
[45] T.Kaczorek, "Local Controllability and Minimum Energy Control of Continuous 2-

Linear Systems with Variable Coefficients", , Multidimensional Systems and Signal Processing, 6, pp.69-75, 1995.


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